

1MHZ CMOS Rail-to-Rail IO Opamp with RF Filter

Features

- Single-Supply Operation from +1.8V ~ +6V
- Rail-to-Rail Input / Output
- Gain-Bandwidth Product: 1MHz (Typ.)
- Low Input Bias Current: 1pA (Typ.)
- Low Offset Voltage: 3.5mV (Max.)
- Quiescent Current: 75µA per Amplifier (Typ.)
- Embedded RF Anti-EMI Filter

- Operating Temperature: -40°C ~ +125°C
- Small Package:

HGV6001 Available in SOT23-5 and SC70-5 Packages
HGV6002 Available in SOP-8 and MSOP-8 Packages
HGV6004 Available in SOP-14 and TSSOP-14 Packages

General Description

The HGV6001 family have a high gain-bandwidth product 1MHz, a slew rate of $0.8V/\mu \, s$, and a quiescent current of $75 \, \mu$ A/amplifier at 5V. The HGV6001 family is designed toprovide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for HGV6001 family. Theyare specified over the extended industrial temperature range (-40° to +125°C). The operating range is from 1.8V to 6V. The HGV6001single is available in Green SC70-5 and SOT23-5 packages. The HGV6002 dual is available in Green SOP-14 and TSSOP-14 packages.

Applications

- ASIC Input or Output Amplifier
- Sensor Interface
- Medical Communication
- Smoke Detectors

- Audio Output
- Piezoelectric Transducer Amplifier
- Medical Instrumentation
- Portable Systems

Pin Configuration

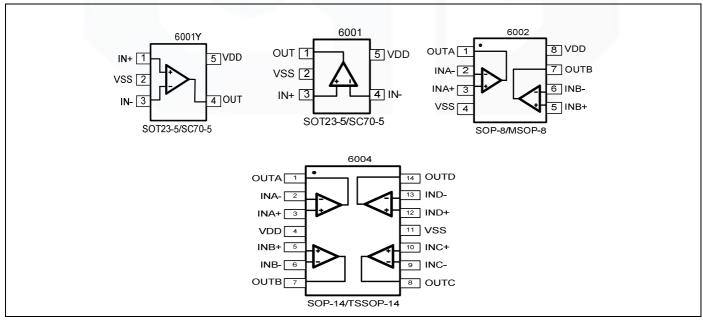


Figure 1. Pin Assignment Diagram



Absolute Maximum Ratings

Condition	Min	Max
Power Supply Voltage (V _{DD} to Vss)	-0.5V	+7.5V
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V _{DD} +0.5V
PDB Input Voltage	Vss-0.5V	+7V
Operating Temperature Range	-40°C	+125°C
Junction Temperature	+160)°C
Storage Temperature Range	-55°C	+150°C
Lead Temperature (soldering, 10sec)	+260)°C
Package Thermal Resistance (T _A =+25℃)	·	
SOP-8, θ _{JA}	125°0	C/W
MSOP-8, θ _{JA}	216°0	C/W
SOT23-5, θ _{JA}	190°0	C/W
SC70-5, θ _{JA}	333°0	C/W
ESD Susceptibility		
НВМ	6K	V
MM	400	V

Note: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

Package/Ordering Information

MODEL	CHANNEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
		HGV6001-CR	SC70-5	Tape and Reel,3000	6001
1101/0004	HGV6001 Single	HGV6001-TR		Tape and Reel,3000	6001
HGV6001		HGS6001Y-CR	SC70-5	Tape and Reel,3000	6001Y
		HGV6001Y-TR	SOT23-5	Tape and Reel,3000	6001Y
HCMeons	Dual	HGV6002-SR	SOP-8	Tape and Reel,4000	HGV6002
HGV6002 Di	Duai	HGV6002-MR	MSOP-8	Tape and Reel,3000	HGV6002
HGV6004	Quad	HGV6004-TR	TSSOP-14	Tape and Reel,3000	HGV6004
		HGV6004-SR	SOP-14	Tape and Reel,2500	HGV6004



Electrical Characteristics

(At VS = +5V, RL = $100k\Omega$ connected to VS/2, and VOUT = VS/2, unless otherwise noted.)

			HGV6001/2/4					
PARAMETER	SYMBOL	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE				
			+25℃	+25℃	-40℃ to +85℃	UNITS	MIN/MAX	
INPUT CHARACTERISTICS								
Input Offset Voltage	Vos	$V_{CM} = V_S/2$	0.8	3.5	5.6	mV	MAX	
Input Bias Current	I _B		1			pA	TYP	
Input Offset Current	los		1			pA	TYP	
Common-Mode Voltage Range	V _{CM}	V _S = 5.5V	-0.1 to +5.6			٧	TYP	
Common Mode Dejection Datio	CMDD	V _S = 5.5V, V _{CM} = -0.1V to 4V	70	62	62	dB		
Common-Mode Rejection Ratio	CMRR	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 5.6V	68	56	55		MIN	
Open Lean Valtage Coin	Δ.	$R_L = 5k\Omega$, $V_O = +0.1V$ to +4.9V	80	70	70	dB		
Open-Loop Voltage Gain	A _{OL}	$R_L = 10k\Omega$, $V_O = +0.1V$ to +4.9V	100	94	85		MIN	
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$	A \	2.7			μV/°C	TYP	
OUTPUT CHARACTERISTICS								
	V _{OH}	R _L = 100kΩ	4.997	4.980	4.970	٧	MIN	
Output Valtage Cuing from Deil	V _{OL}	R _L = 100kΩ	5	20	30	mV	MAX	
Output Voltage Swing from Rail	V _{OH}	R _L = 10kΩ	4.992	4.970	4.960	٧	MIN	
	V _{OL}	R _L = 10kΩ	8	30	40	mV	MAX	
Output Current	I _{SOURCE}	D = 100 to 1/ /2	84	60	45	A	MIN	
Output Current	I _{SINK}	$R_L = 10\Omega$ to $V_S/2$	75	60	45	mA	IVIIIN	
POWER SUPPLY								
Operating Voltage Dange				1.8	1.8	٧	MIN	
Operating Voltage Range				6	6	٧	MAX	
Power Supply Rejection Ratio	PSRR	$V_S = +2.5V \text{ to } +6V, V_{CM} = +0.5V$	82	60	58	dB	MIN	
Quiescent Current / Amplifier	ΙQ		75	110	125	μA	MAX	
DYNAMIC PERFORMANCE (CL	= 100pF)							
Gain-Bandwidth Product	GBP		1			MHz	TYP	
Slew Rate	SR	G = +1, 2V Output Step	0.8			V/µs	TYP	
Settling Time to 0.1%	t _S	G = +1, 2V Output Step	5.3			μs	TYP	
Overload Recovery Time		V _{IN} ·Gain = V _S	2.6			μs	TYP	
NOISE PERFORMANCE								
Voltago Noigo Dansity	_	f = 1kHz	27			nV/\sqrt{Hz}	TYP	
Voltage Noise Density	e _n	f = 10kHz	20			nV/\sqrt{Hz}	TYP	



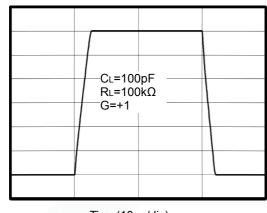
Output Voltage (250mV/div)

Typical Performance characteristics

At T_A =+25°C, Vs=5V, R_L =100K Ω connected to V_S /2 and V_{OUT} = V_S /2, unless otherwise noted.

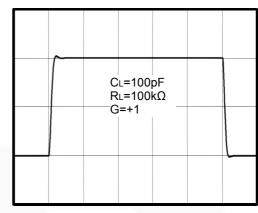
Output Voltage (50mV/div)





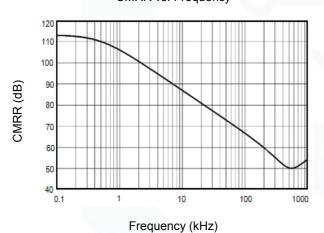
Time(10µs/div)

Small Signal Transient Response

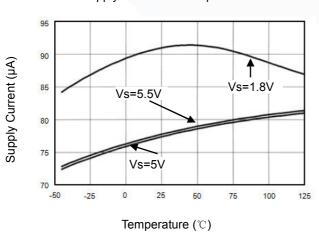


Time(2µs/div)

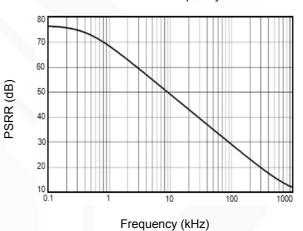
CMRR vs. Frequency



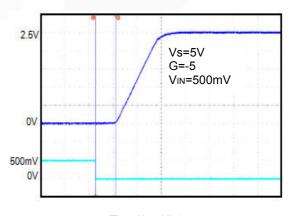
Supply Current vs. Temperature



PSRR vs. Frequency



Overload Recovery Time



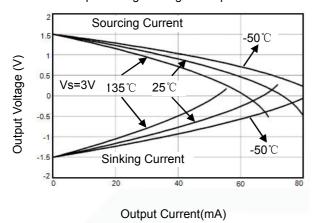
Time(2µs/div)



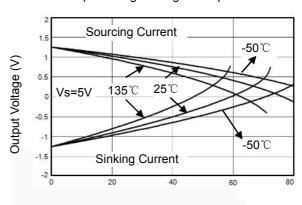
Typical Performance characteristics

At T_A =+25°C, R_L =100K Ω connected to V_S /2 and V_{OUT} = V_S /2, unless otherwise noted.

Output Voltage Swing vs.Output Current

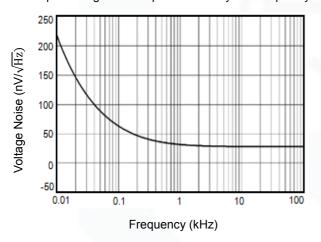


Output Voltage Swing vs.Output Current

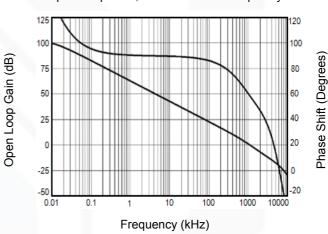


Output Current(mA)

Input Voltage Noise Spectral Density vs. Frequency



Open Loop Gain, Phase Shift vs. Frequency





Application Note

Size

HGV6001 family series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the HGV6001 family packages save spaceon printed circuit boards and enable the design of smaller electronic products.

Power Supply Bypassing and Board Layout

HGV6001 family series operates from a single 1.8V to $\mbox{\sc W}$ supply or dual $\pm 0.9 \mbox{\sc W}$ supplies. For best performance, a $0.1 \mu \mbox{\sc F}$ ceramic capacitor should be placed close to the $\mbox{\sc V}_{DD}$ pin in single supply operation. For dual supply operation, both $\mbox{\sc V}_{DD}$ and $\mbox{\sc V}_{SS}$ supplies should be bypassed to ground with separate $0.1 \mu \mbox{\sc F}$ ceramic capacitors.

Low Supply Current

The low supply current (typical 75µA per channel) of HGV6001 family will help to maximize battery lifeThey are ideal for battery powered systems

Operating Voltage

HGV6001 family operates under wide input supply voltage (1.8V to 6V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-lon battery lifetime

Rail-to-Rail Input

The input common-mode range of HGV6001 family extends 00mV beyond the supply rails (V_{SS} -0.1V to V_{DD} +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of HGV6001 family can typically swing to less than 0mV from supply rail in light resistive loads (>100k Ω), and 60mV of supply rail in moderate resistive loads (10k Ω).

Capacitive Load Tolerance

The HGV6001 family is optimized for bandwidth and sped, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2 shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

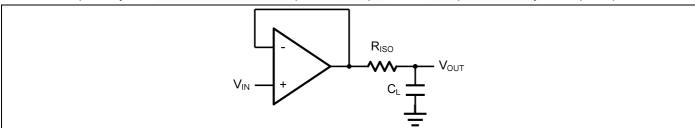


Figure 2 Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the R_{ISO} resistor value, the more stable V_{OUT} will be. However, if there is a resistive load R_L in parallel with the capacitive load, a voltage divider (proportional to R_{ISO}/R_L) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2. R_F provides the DC accuracy by feed-forward the V_{IN} to R_L. C_F



and $R_{\rm ISO}$ serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of C_F . This in turn will slow down the pulse response.

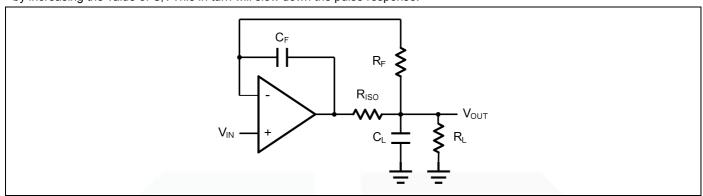


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy



Typical Application Circuits

Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using HGV6001 family

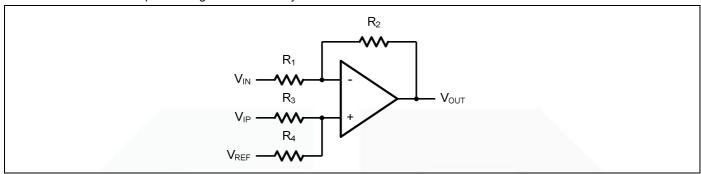


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R₁=R₃ and R₂=R₄), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by $-R_2/R_1$. The filter has a -20dB/decade roll-off after its corner frequency $f_C=1/(2\pi R_3C_1)$.

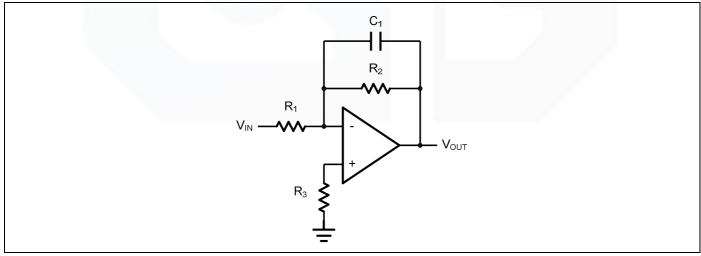


Figure 5. Low Pass Active Filter



Instrumentation Amplifier

The triple HGV6001 family can be used to build a thre-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R_2/R_1 . The two differential voltage followers assure the high input impedance of the amplifier.

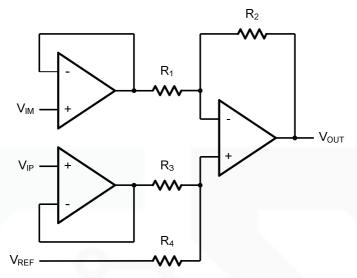
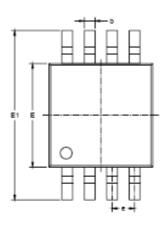


Figure 6. Instrument Amplifier

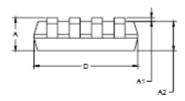


Package Information

MSOP-8



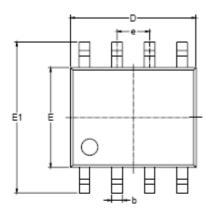


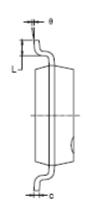


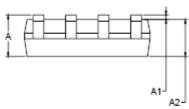
Symbol	Dimen In Milli		Dimensions In Inches		
	MIN	MAX	MIN	MAX	
A	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.006	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
e	0.650 BSC		0.026 BSC		
L	0.400	0.800	0.016	0.031	
θ	0° 6° 0°		6°		



SOP-8



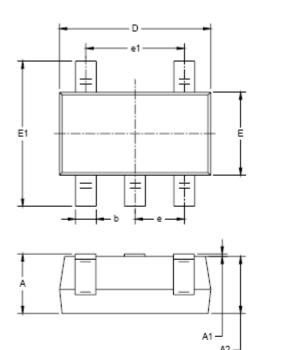


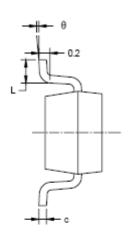


Symbol		nsions imeters	Dimensions In Inches		
•	MIN	MAX	MIN	MAX	
Α	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
e	1.27 BSC		0.050	BSC	
L	0.400	1.270	0.016	0.050	
е	0°	8°	0°	8°	



SOT23-5

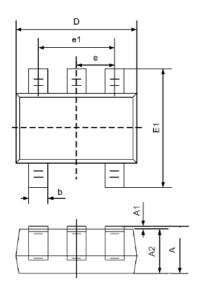


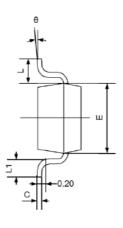


Symbol		nsions imeters	Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
Α	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
E	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
e	0.950) BSC	0.037	BSC	
e1	1.900	1.900 BSC		BSC	
L	0.300	0.600	0.012	0.024	
Θ	0°	8°	0° 8°		



SC70-5

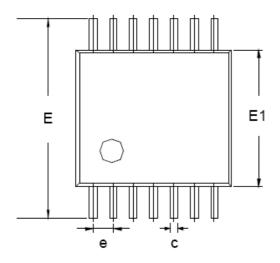


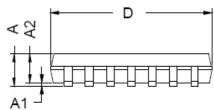


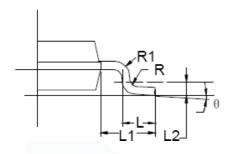
	Dimens	sions	Dimensions		
Symbol	In Milli	meters	In Inches		
	Min	Max	Min	Max	
Α	0.900	1.100	0.035	0.043	
A1	0.000	0.100	0.000	0.004	
A2	0.900	1.000	0.035	0.039	
b	0.150	0.350	0.006	0.014	
С	0.080	0.150	0.003	0.006	
D	2.000	2.200	0.079	0.087	
E	1.150	1.350	0.045	0.053	
E1	2.150	2.450	0.085	0.096	
е	0.650T	ΥP	0.026TYP		
e1	1.200	1.400	0.047	0.055	
L	0.525REF		0.021REF		
L1	0.260	0.460	0.010	0.018	
θ	0°	8°	0°	8°	



TSSOP-14



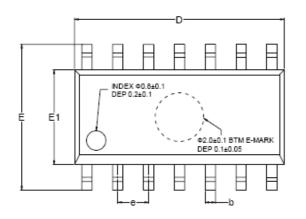


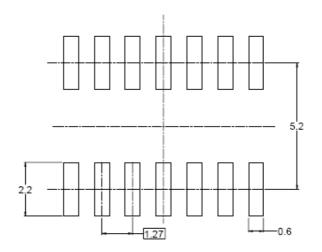


	Dimensions				
Sumb at	In Millimeters				
Symbol	MIN	TYP	MAX		
А	-	-	1.20		
A1	0.05	-	0.15		
A2	0.90	1.00	1.05		
b	0.20	-	0.28		
С	0.10	-	0.19		
D	4.86	4.96	5.06		
E	6.20	6.40	6.60		
E1	4.30	4.40	4.50		
е		0.65 BSC			
L	0.45	0.60	0.75		
L1	1.00 REF				
L2	0.25 BSC				
R	0.09	-	-		
θ	0°	-	8°		

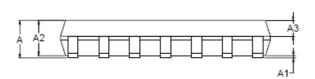


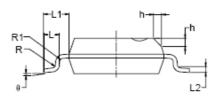
SOP-14





RECOMMENDED LAND PATTERN (Unit: mm)





Comple ed	Dimensions In Millimeters			Dimensions In Inches		
Symbol	MIN	MOD	MAX	MIN	MOD	MAX
Α	1.35		1.75	0.053		0.069
A1	0.10		0.25	0.004		0.010
A2	1.25		1.65	0.049		0.065
A3	0.55		0.75	0.022		0.030
b	0.36		0.49	0.014		0.019
D	8.53		8.73	0.336		0.344
Е	5.80		6.20	0.228		0.244
E1	3.80		4.00	0.150		0.157
е	1.27 BSC				0.050 BSC	
L	0.45		0.80	0.018		0.032
L1	1.04 REF				0.040 REF	
L2	0.25 BSC			0.01 BSC		
R	0.07			0.003		
R1	0.07			0.003		
h	0.30		0.50	0.012		0.020
θ	0°		8°	0°		8°